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UTILITY PATENT APPLICATION TRANSMITTAL

(Only for new nonprovisional applications under 37 C.F.R. § 1.53(b))

Attorney Docket No. **367.38669X00**

First Inventor or Application Identifier **Simon FURMIDGE**

Title **See 1 in Addendum**

Express Mail Label No.

APPLICATION ELEMENTS

See MPEP chapter 600 concerning utility patent application contents.

1. ☒ * Fee Transmittal Form (e.g., PTO/SB/17)
(Submit an original and a duplicate for fee processing)
2. ☒ Specification [Total Pages **11**]
(preferred arrangement set forth below)
 - Descriptive title of the Invention
 - Cross References to Related Applications
 - Statement Regarding Fed sponsored R & D
 - Reference to Microfiche Appendix
 - Background of the Invention
 - Brief Summary of the Invention
 - Brief Description of the Drawings (if filed)
 - Detailed Description
 - Claim(s)
 - Abstract of the Disclosure
3. ☒ Drawing(s) (35 U.S.C. 113) [Total Sheets **2**]
4. Oath or Declaration [Total Pages **2**]
 - a. ☒ Newly executed (original or copy)
 - b. ☐ Copy from a prior application (37 C.F.R. § 1.63(d))
(for continuation/divisional with Box 16 completed)
 - i. ☐ DELETION OF INVENTOR(S)
Signed statement attached deleting inventor(s) named in the prior application, see 37 C.F.R. §§ 1.63(d)(2) and 1.33(b).

NOTE FOR ITEMS 1 & 13 IN ORDER TO BE ENTITLED TO PAY SMALL ENTITY FEES, A SMALL ENTITY STATEMENT IS REQUIRED (37 C.F.R. § 1.27). EXCEPT IF ONE FILED IN A PRIOR APPLICATION IS RELIED UPON (37 C.F.R. § 1.28).

ADDRESS TO:

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5. ☐ Microfiche Computer Program (Appendix)
6. Nucleotide and/or Amino Acid Sequence Submission (if applicable, all necessary)
 - a. ☐ Computer Readable Copy
 - b. ☐ Paper Copy (identical to computer copy)
 - c. ☐ Statement verifying identity of above copies

ACCOMPANYING APPLICATION PARTS

7. ☒ Assignment Papers (cover sheet & document(s))
8. ☐ 37 C.F.R. § 3.73(b) Statement of (when there is an assignee) ☒ Attorney
9. ☐ English Translation Document (if applicable)
10. ☒ Information Disclosure Statement (IDS)/PTO-1449 ☒ Copies of IDS Citations
11. ☐ Preliminary Amendment
12. ☒ Return Receipt Postcard (MPEP 503)
(Should be specifically itemized)
13. ☐ * Small Entity Statement(s) ☐ Statement filed in prior application Status still proper and desired (PTO/SB/08-12)
14. ☐ Certified Copy of Priority Document(s) (if foreign priority is claimed)
15. ☒ Other: **Figs. 1-2**

16. If a CONTINUING APPLICATION, check appropriate box, and supply the requisite information below and in a preliminary amendment:
☐ Continuation ☐ Divisional ☐ Continuation-in-part (CIP) of prior application No. _____
 Group / Art Unit: _____

Prior application information: Examiner _____
 For CONTINUATION or DIVISIONAL APPS only: The entire disclosure of the prior application, from which an oath or declaration is supplied under Box 4b, is considered a part of the disclosure of the accompanying continuation or divisional application and is hereby incorporated by reference. The incorporation can only be relied upon when a portion has been inadvertently omitted from the submitted application parts.

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Name (Print/Type) Carl J. Brundige	Registration No. (Attorney/Agent) 29,621
Signature <i>[Signature]</i>	Date June 23, 2000

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IN THE UNITED STATES PATENT AND TRADEMARK OFFICE

Applicants: Simon FURMIDGE
Serial No.: Not yet assigned
Filed: June 23, 2000
For: TRANSMITTER AND A MODULATOR THEREFOR
Group: Not yet assigned
Examiner: Not yet assigned

PRELIMINARY AMENDMENT

Assistant Commissioner for Patents
Washington, D.C. 20231

June 23, 2000

Sir:

Prior to examination, please amend the above-identified application as follows.

IN THE SPECIFICATION

Please amend the specification as follows:

Page 9, line 11, delete "What is claimed is:-".

IN THE CLAIMS

Page 10, before line 1, insert --What is claimed is:--.

IN THE ABSTRACT

Line 12, delete "Figure 1".

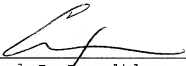
REMARKS

Entry of the above amendments is respectfully requested.

Please charge any shortage in fees due in connection with the filing of this paper, or credit any overpayment of fees, to the deposit account of Antonelli, Terry, Stout & Kraus, LLP, Deposit Account No. 01-2135 (367.38669X00).

Respectfully submitted,

ANTONELLI, TERRY, STOUT & KRAUS, LLP



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A TRANSMITTER AND A MODULATOR THEREFOR**5 Background of the Invention**

The present invention relates to a transmitter for a portable radio communication apparatus, and more particularly to a modulator for a direct conversion transmitter.

10

A general trend in portable communication apparatus is the reduction in volume, weight and power consumption of such apparatus. This has led to efforts towards reducing the number of elements and devices necessary to perform the functions associated with portable communication devices. In particular, the radio frequency transmit strip of portable communication apparatus, which typically comprises a number of up-converting stages, is an area in which a reduction in the number of elements and devices would be beneficial.

20 One approach to reduce the number of stages in the radio frequency transmit strip is to convert a baseband signal (comprising the information to be transmitted) to a radio frequency carrier signal in a single step. This is termed direct conversion or direct modulation. To carry out direct conversion, a local oscillator signal (LO) having the same frequency as the required radio
25 frequency carrier signal is mixed with the baseband signal in a suitable non-linear device such as a mixer diode. The output of the mixer contains the sum and difference of the LO and the baseband signal. In this way, the LO signal is modulated by the baseband signal.

30 Typically, the baseband signal comprises 'I' and 'Q' components and accordingly two such modulators with their outputs summed together are

required. These are fed with two LO signals, the 'Q' component LO having a 90 degree phase shift with respect to the 'I' component.

Therefore, one of the issues with direct conversion transmitter design is that of generating the 90 phase shift for the 'Q' LO signal. At present, there are two commonly used methods: one is a passive phase shift network using reactive elements, the other is an active divide-by-two circuit. The passive phase shift network approach involves the design of a frequency selective 'All Pass' filter network to provide the phase difference between the 'I' and 'Q' local oscillators but has the disadvantage that it is inherently limited to substantially narrow band applications and also that it can be difficult to integrate such networks onto an IC. The divide-by-two approach uses a pair of dividers one clocked off the rising and one off the falling edge of an LO at double the wanted operating frequency. The active divide-by-two circuit has the disadvantage that it requires high current to operate and requires an LO of double the frequency of operation. For example, if the wanted operating frequency were say 1.9GHz then an LO of 3.8GHz would be required to drive this circuit. Hence, a synthesizer and divide-by-two operating at 3.8GHz would consume a great deal of current. Since the trend in portable communication devices is towards compactness there is less space for battery packs, and thus less battery capacity is available and so low current operation becomes increasingly important to achieve acceptable talk and standby times.

A separate issue, but one that is also important in the design of a direct conversion transmitter, is that of the gain control of the transmitter. In certain radio telephone systems such as WCDMA the transmitter is required to vary its output power over a very wide range e.g. 70 dB for WCDMA. Additionally, a direct conversion transmitter must obtain all of its control range at radio frequencies. Thus, since the LO is operating at the wanted radio frequency, problems of LO leakage and shielding can become very significant.

Conventionally, with one or more IFs, variable gain amplifiers and/or attenuators are distributed between the RF and intermediate frequency (baseband) and used to vary the transmitter power. These approaches however have not satisfactorily provided the wide range of power control required.

A further concern with direct conversion transmitters is that the LO signal cannot be provided directly from a synthesiser locked VCO. There are two main reasons why. Firstly, if the radiotelephone has an internal antenna there is a very great risk that the transmitter will radiate back into the synthesiser locked VCO and cause it to go out of lock or generate spurious. Secondly, there will be insufficient isolation between antenna impedance (which will vary a great deal as the user moves around) and the synthesiser locked VCO. This will cause the synthesiser locked VCO to either go out of lock or generate spurious. One method of solving this problem is to create the LO signal by mixing together two synthesiser locked VCO signals and then filtering the LO to remove any unwanted mixing products. This however, increases component count and current consumption.

Summary of the Invention

Against this background, and in one aspect, the present invention provides a transmitter for a portable radio communication apparatus comprising a modulator having first port for inputting a baseband signal and a second port for inputting a local oscillator signal, and including means for rectifying the input local oscillator signal to provide a conductance waveform at a multiple of the local oscillator signal, and means for mixing the baseband signal with the conductance waveform at said multiple of the local oscillator signal frequency for up-converting the baseband signal to a radio frequency modulated carrier, the transmitter including means for controlling the gain of the modulator thereby to control the output level of the modulator.

By means of the invention, the baseband signal can be up-converted without employing a local oscillator at the carrier frequency of the transmitted signal, and the output of the modulator can be varied.

- 5 The local oscillator can be fixed to operate at a sub-harmonic of the transmitted signal frequency, and the mixing action is performed between the baseband signal and one of the harmonics of the local oscillator, which advantageously ensures that the harmonic is not generated in the transmitter. This is termed sub-harmonic mixing. Accordingly, the local oscillator signal
10 frequency is far removed from transmitted signal frequency, resulting in the mixer providing very high local oscillation to transmitted signal isolation.

- A further advantage is that the high local oscillator isolation allows the output level of the modulator to be controlled over a large range before local
15 oscillator breakthrough becomes a problem.

The modulator output may be controlled by means of controlling the current through the modulator circuit.

- 20 The LO signal advantageously requires only a single ended drive, resulting in lower current consumption and component count.

- In a preferred embodiment of the present invention, the modulator comprises two cross-coupled pairs of switching elements in the form of two cross
25 connected long tail pairs of bipolar transistors, wherein a signal input modulates the switching elements at a multiple of the local oscillator frequency.

- Advantageously, the switching elements use substantially identical devices
30 and in a preferred arrangement provide a high degree of balance. This further improves isolation between the radio frequency and local oscillator ports,

allowing still greater control range of the output level. The use of substantially identical switching devices facilitates their integration into an integrated circuit.

5 For this application of direct conversion transmitter the local oscillator is half the transmitted frequency.

10 The present invention extends to apparatus for a sub-harmonic mixer, comprising switching means, a first port for inputting a baseband signal to the switching means to be up-converted, and a second port for inputting a local oscillator signal to drive the switching means at an even multiple of the local oscillator frequency for up-converting the baseband signal to transmission frequency.

15 The mixer further includes means for controlling the output level of the up-converted the baseband signal.

An advantageous feature of the mixer disclosed herein is that it is bi-directional.

20 The invention further includes methods for performing the modulating defined herein.

Brief Description of the Drawings

25 Specific embodiments in accordance with the invention are hereinafter described, by way of example only, and with reference to the accompanying drawings, in which:

30 Figure 1 is a schematic circuit diagram of a preferred embodiment in accordance with the present invention; and

Figure 2 schematically shows the voltage waveforms of the local oscillator and first signal input ports of the embodiment of Figure 1.

Detailed Description of the Invention

5

Referring initially to Figure 1, there is shown an exemplary modulator circuit diagram of the present invention. In this circuit a suitable power supply is connected at V3v ie. 3 volts. The power supply V3v is connected to the output loads of the modulator circuit, the load being represented by resistors
 10 RL1 and RL2. These however may be reactive, resistive or active whichever is most suitable in the application. The output loads RL1 and RL2 are connected respectively to Pout and Nout which are the differential outputs of the modulator circuit. The signal present between Pout and Nout is the modulated radio frequency carrier. In the case of an 'I/Q' modulator this
 15 differential output would be summed with the differential output of a second modulator. The summed output would then be passed to the next stage of the transmitter.

Pout and Nout are coupled to the transistor network forming the switching
 20 means of the modulator circuit. The transistors Q1,2,3,4 are shown as bipolar junction transistors, but could be FETs or the like. Q1 and Q2 form a long tail pair as do Q3 and Q4. The collectors of Q1 and Q3 are connected to Pout, and the collectors of Q2 and Q4 are connected to Nout. RFP is connected to the base of Q1 and RFN is connected to the base of Q4; RFP and RFN
 25 provide the differential inputs to the switching means, whereby the I or Q baseband signals can be injected differentially into these ports.

Lo designates the local oscillator input port and this is connected to the bases of Q2 and Q3. This will cause the switching means to operate at twice the
 30 frequency of input.

The emitters of each pair of long tail pairs of transistor network Q1 to Q4 are connected to a gain control means which controls the current through the transistor network. The gain control means comprises two current controllers labelled Current 1 and Current 2, and the level labelled Gain is an input to the current controllers whereby the current in both the pairs of long tail pairs may be increased or decreased. Vneg may be connected to ground or a suitable positive or negative power supply.

The operation of the circuit will now be described.

The current controller operates to hold the total current through the Q1,Q2 and Q3,Q4 pairs constant. If the voltage on RFP and RFN is held constant for instance, the local oscillator signal on Lo as it becomes more positive will increase the current flow in Q2 and Q3 thus reducing the current flow in Q1 and Q4. When the local oscillator signal is lower in voltage than RFP and RFN, Q2 and Q3 will have much less current flowing in them than Q1 and Q4. When the local oscillator voltage is the same as RFP, Q1 and Q2 and also Q3 and Q4 will have the same current flowing in them. When the local oscillator voltage is greater than the voltage on RFP and RFN Q2 and Q3 will have more current flowing in them than Q3 and Q4.

As shown in the embodiment of Figure 1, the collectors of Q1 and Q3, Q2 and Q4 are connected together. Based on the description above, with RFP and RFN held at the same constant voltage, the current in Q1/Q4 will decrease at the same rate as the current in Q2/Q3 increases as the Lo voltage increases. Thus the voltage at Pout and Nout will remain constant. If the Local oscillator signal on Lo is larger in amplitude than Vbe i.e. 0.7 volts, it will completely switch Q2 and Q3 on and off. This is the desired mode of operation. The output pulses will then be of an amplitude defined solely by the amplitude of the differential signal applied between RFP and RFN and the gain defined by the current control means.

If a constant offset voltage is applied between RFP and RFN it will change the point on the Lo waveform where Q3 switches on relative to Q2. As Q3's collector is cross coupled to Q1's collector, if Q2 switches later than Q3 then the current flow in Q3 will increase before the current flow in Q1 reduces.

5 Thus an increase in current flow will be seen in RL1 and because Q4's collector is coupled to Q2s a corresponding reduction in current flow in RL2 will be seen. This will result in a voltage pulse between Pout and Nout. This effect occurs on both the rising and falling edges of the local oscillator signal. Thus the mixing effect takes place at twice the local oscillator frequency.

10

Referring to Figure 2, this shows the operation of the sub-harmonic mixer with an input signal of 4.1MHz with an Lo of 2Mhz. This demonstrates its mode of operation.

15 As can be seen from the Lo and output waveforms, the output pulses occur, in this case, at the zero crossing points of the Lo. The output pulses if referenced to the first input signal have a amplitude that is proportional to the amplitude of the first input signal at that point in time. The ratio of the first input signal amplitude to the amplitude of the Output voltage (Gain) is
20 determined by the current flow in the Q1,2 and Q2,3 long-tail pairs. This current flow is determined by the current control means in the emitters of the Q1,2 and Q3,4 long tail pairs.

The present invention may be embodied in other specific forms without
25 departing from its essential attributes. The switching means shown as bipolar transistors may be replaced with suitable FETs or other forms of voltage or current controlled switches. Accordingly reference should be made to the appended claims and other general statements herein rather than to the foregoing specific description as indicating the scope of invention.

30

Furthermore, each feature disclosed in this specification (which term includes the claims) and/or shown in the drawings may be incorporated in the invention independently of other disclosed and/or illustrated features. In this regard, the invention includes any novel features or combination of features disclosed
5 herein either explicitly or any generalisation thereof irrespective of whether or not it relates to the claimed invention or mitigates any or all of the problems addressed.

The appended abstract as filed herewith is included in the specification by
10 reference.

What is claimed is:-

1. A transmitter for a portable radio communication apparatus comprising a modulator having a first port for inputting a baseband signal and a second port for inputting a local oscillator signal, and including means for rectifying the input local oscillator signal to provide a conductance waveform at a multiple of the local oscillator signal, and means for mixing the baseband signal with the conductance waveform at said multiple of the local oscillator signal frequency for up-converting the baseband signal to a radio frequency modulated carrier, the transmitter including means for controlling the gain of the modulator thereby to control the output level of the modulator.
2. A transmitter according to claim 1, wherein a local oscillator signal drives the switching means at a multiple of its frequency.
3. A transmitter according to claim 1, wherein the means for controlling the gain of the modulator comprises current control means.
4. A transmitter according to claim 1, wherein the modulator comprises two cross-coupled pairs of switching elements, wherein a signal input modulates the switching elements at a multiple of the local oscillator frequency.
5. A transmitter according to claim 4, wherein said two cross-coupled pairs of switching elements comprise two cross connected long tail pairs of bipolar transistors.
6. Apparatus for a sub-harmonic mixer, comprising switching means, a first port for inputting a baseband signal to the switching means to be up-converted, and a second port for inputting a local oscillator signal to drive the switching means at an even multiple of the local oscillator frequency for up-converting the baseband signal to transmission frequency.

ABSTRACT

A transmitter for a portable radio communication apparatus comprising a modulator having first port for inputting a baseband signal and a second port
5 for inputting a local oscillator signal, and including means for rectifying the input local oscillator signal to provide a conductance waveform at a multiple of the local oscillator signal, and means for mixing the baseband signal with the conductance waveform at said multiple of the local oscillator signal frequency for up-converting the baseband signal to a radio frequency modulated carrier,
10 the transmitter including means for controlling the gain of the modulator thereby to control the output level of the modulator.

Figure 1

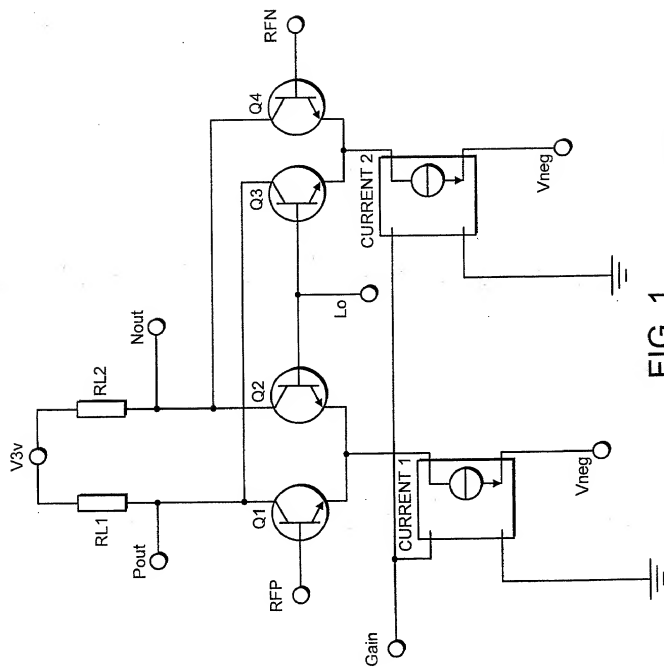


FIG. 1

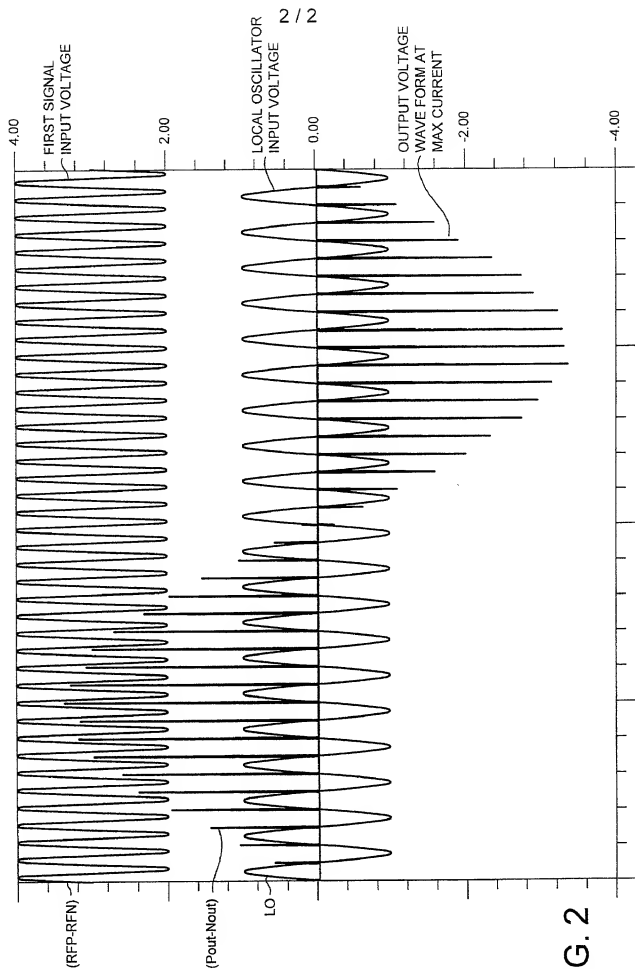


FIG. 2

DECLARATION AND POWER OF ATTORNEY - PATENT APPLICATION

As a below named inventor, I hereby declare: that my citizenship, residence and post office address are as stated below; that I verily believe I am the original, first and sole inventor (if only one is named below) or a joint inventor (if plural inventors are named below) of the invention entitled: A TRANSMITTER AND A MODULATOR THEREFOR

the specification of which _____ is attached hereto

_____ was filed on _____ as Application
Serial No. _____ and was amended on
_____.

I hereby state that I have reviewed and understand the contents of the above-identified specification, including the claims, as amended by any amendment referred to above. I acknowledge the duty to disclose information which is material to the examination of this application in accordance with Title 37, Code of Federal Regulations, §1.56(a). I hereby claim foreign priority benefits under Title 35, United States Code, §119 of any foreign application(s) for patent or inventor's certificate listed below and have also identified below any foreign application for patent or inventor's certificate having a filing date before that of the application on which priority is claimed:

Prior Foreign Application(s)			Priority Claimed
Number	Country	Day/Month/Year Filed	Yes / No
9914750.6	GREAT BRITAIN	24 JUNE 1999	YES

I hereby claim the benefit under Title 35, United States Code, §120 of any United States application(s) listed below and, insofar as the subject matter of each of the claims of this application is not disclosed in the prior United States application in the manner provided by the first paragraph of Title 35, United States Code, §112, I acknowledge the duty to disclose material information as defined in Title 37, Code of Federal Regulations, §1.56(a) which occurred between the filing date of the prior application and the national or PCT international filing date of this application;

Application Serial No	Status-patented, pending or abandoned
-----------------------	---------------------------------------

I hereby appoint as principal attorneys: Donald R. Antonelli, Reg. No. 20,296; David T. Terry, Reg. No. 20,178; Melvin Kraus, Reg. No. 22,466; William I. Solomon, Reg. No. 28,565; Gregory E. Montone, Reg. No. 28,141; Ronald J. Shore, Reg. No. 28,577; Donald E. Stout, Reg. No. 26,422; Alan E. Schiavelli, Reg. No. 32,087; James N. Dresser, Reg. No. 22,973, Carl I. Brundidge, Reg. No. 29,621; and Paul J. Skwierawski, Reg. No. 32,173; to prosecute and transact all business in the Patent and Trademark Office connected with this

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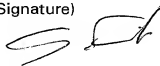
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